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Research Paper

Analysis of VSWR and Coupling of a Longitudinal Slot Coupled Shunt Tee Junction for Non-Standard Rectangular Waveguide

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ABSTRACT:- The waveguide junctions are used to split waves from one waveguide into other waveguides. In order to make a three-port waveguide junction, it is possible to connect a third arm in a waveguide. The waveguide can be added across the long dimension of the waveguide or across the narrow dimension of the waveguide. The later connection yields a H-plane Tee. A H-plane Tee is electrical equivalent of connecting the arm in parallel or in shunt. In this junction, the circulating magnetic field traveling down in arm 3 meets the junction between arm 1 and 2 and splits the power between the two arms if, as is common, they are of same length. The resulting H-field has the same phase and magnitude in each arm. In this present paper, such longitudinal slots are considered to couple power from feed guide to the coupled guide. However, the data on the variation of admittance, coupling and VSWR are computed for slots of resonant length and other slot parameters The additional feature of this work contains the use of non-standard waveguides as they provide additional parameters for the design of the junctions as well as array of such junctions.

Keywords:- Waveguides, Shunt Tees, Admittance Loading, H-Plane Tee, Slot coupled Tee Junctions

I. INTRODUCTION

A narrow slot cut in the wall of a rectangular waveguide is an useful radiating element in highly directive radar antennas and flush mounted antennas. The radiated power in a specified direction is controlled by the slot parameters, dimensions of the waveguide and geometry of the slot including its location, feed network and exciting function. Slots cut in the narrow wall should be inclined at least by a small angle from the vertical axis as the vertical slot does not radiate. Earlier papers dealing with both resonant and non-resonant slots used equivalent circuits for shunt and series slots which were constructed by employing a standing wave formulation of the waveguide modes. The analysis of H-plane Tee junction containing standard rectangular waveguides is extensively reported [1-3]. In these junctions, coupling takes place through a rectangular slot from primary to secondary guide. Moreover, such structures cannot accommodate the slots of required dimensions due to the fixed dimension of waveguides. In order to have flexibility in the design, non-standard waveguides are proposed for the fabrication of H-plane Tee junctions.

The variational expressions are used to obtain slot admittance parameters by some researchers. The susceptances are obtained from stored energy considerations. It may be noted that a slot couples the energy to the Tee arm and the Tee arm in turn radiates into free space. The slot of present interest has rectangular ends. The data on susceptance parameter is used to obtain S- parameters and hence coupling and VSWR. The variations of admittance, coupling and VSWR are obtained as a function of frequency as well as slot and waveguide dimensions. The data presented in this work is unique and is extremely useful for the design of array of H-plane Tee junction radiators. The variational approach is found to be highly involved and it is difficult to bring out equivalent circuit parameter accurately. The equivalent network parameter as reported in the literature [94] is found to have a shunt element which consists of a constant real part and a variable imaginary part. This cannot be true as the conductance and the susceptance of the admittance parameter vary in coupling and VSWR. The analysis of the junction is carried out using hybrid modes as well as TE and TM concepts. The concept of

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self reaction proposed by Rumsey [4] and Harrington [5] is used to derive admittance parameter. The evaluation of equivalent shunt parameter takes care of energy storage in the feed guide. The self reactions involved in both the guides are evaluated along with discontinuity in modal current.

Self reaction in the primary guide is obtained from vector potential. This potential is obtained from the solution of Helmholtz equation. The hybrid mode concept is used to obtain the self reaction in the primary guide and TE and TM mode concept is used to obtain self reaction in the coupled guide.

II. ANALYSIS

Consider H-plane Tee junction of fig. 1 for the analysis.

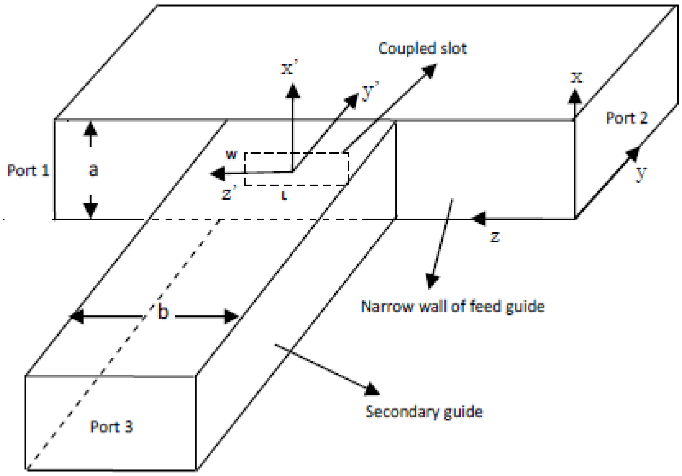

Fig. 1 ***H-plane Tee junction using longitudinal slot***

Here, the feed waveguide and coupled guide have non-standard dimensions to have more flexibility in the design of the junction and also in the design of the junction radiators. The prime objective of the present work is to use the junction as the radiator of vertically polarized fields. Assume the electric field as sinusoidal and it is replaced by its equivalent magnetic current for the junction analysis. The admittance loading is evaluated using the concept of the magnetic current and the modal voltages of TE and TM modes are given by [5]. Arms containing port 1 and 2 are in shunt when input is given to port 3. It is therefore called a shunt Tee. A H-plane Tee junction can be used as a radiator, when the Tee arm is coupled to the primary guide through a slot. The radiating field from the Tee arm is always vertically polarized irrespective of the slot geometry and orientation. The H-plane Tee junction coupled through a longitudinal slot is analyzed by a few researchers.
The slot admittance is represented by

$$Y_s = \frac{\sum_m \sum_n \left[{V_{mn}^{e}}^2 Y_o^e + {V_{mn}^{m}}^2 Y_o^m \right]}{V_{10}^e . V_{10}^e}$$

Here Y_o^e and Y_o^m are the wave admittances of TE and TM modes. And

$$V_{mn}^e = \int_{-W/2}^{W/2} \int_{-L/2}^{L/2} \mathbf{E}_{slot} . \mathbf{e}_{mn}^e \, dx' dz'$$

$$V_{mn}^m = \int_{-W/2}^{W/2} \int_{-L/2}^{L/2} \mathbf{E}_{slot} . \mathbf{e}_{mn}^m \, dx' dz'$$

The electric field in the slot is given by

$$E_{slot} = \begin{cases} E_s \cos\frac{\pi z'}{L} & -\frac{L}{2} < z' < \frac{L}{2} \\ & -\frac{W}{2} < x' < \frac{W}{2} \\ 0 & \text{elsewhere} \end{cases}$$

The reaction of the fields on their own sources is known as self reaction. The expression for self-reaction of fields ($\mathbf{E}_a$,$\mathbf{H}_a$) on their own sources ($\mathbf{J}_a$,$\mathbf{M}_a$) is given by

$$\langle a, a \rangle = \int_v (\mathbf{E}_a.\mathbf{J}_a - \mathbf{H}_a.\mathbf{M}_a)\, dv$$

The electromagnetic fields propagate freely in free space. However, when they are confined to continuous metallic waveguide structure, the propagation takes place with multiple reflections from the walls of the waveguide. If the walls of the waveguide are perfectly conducting, the fields cannot penetrate into walls, as neither **E** nor **H** exists inside a perfect conductor. On the other hand, the propagation characteristics of electromagnetic fields are completely modified in the presence of discontinuities. The properties of discontinuities are easily represented by equivalent networks.

The overall behavior of the fields in waveguides depends on the type of discontinuity as well as on the field propagation in the waveguides. In the equivalent networks, voltages and currents are introduced to represent the amplitudes of the fields at the terminals of the discontinuity and at all locations in waveguides. Such voltages and currents characterize the power flow into and out of discontinuity but not the fields in the immediate vicinity.For convenience, a field representation is made use of to obtain the relation between the terminal voltages and currents in the metallic structures containing discontinuities.

A longitudinal or an inclined slot in the narrow or broad wall of a rectangular waveguide produces a discontinuity in modal current. Marcuvitz and Schwinger [6] established the basic formula for the discontinuity in modal current and it is given by

$$I = -2jY_{01}WE_{fm}\sin\theta \frac{\sin\left(\frac{W}{2}\beta_{01}\cos\theta\right)}{\left(\frac{W}{2}\beta_{01}\cos\theta\right)}\left(\frac{2}{ab}\right)^{1/2}\frac{\pi}{b\beta_{01}}\cdot\frac{K}{(\beta_{01}\sin\theta)^2 - K^2}$$
$$\left[\cos\left(\beta_{01}\frac{L}{2}\sin\theta\right) - \cos\left(K\frac{L}{2}\right)\right]$$

Where, Y_{01} and b_{01} are the characteristic wave admittance and propagation constant, $\mathbf{h_{01}}$ and $\mathbf{h_{z01}}$ are the transverse and longitudinal modal vector functions of the magnetic field for dominant mode in feed guide

The junction provides vertically polarized fields irrespective of the slot orientation. The admittance loading is evaluated in the present work is obtained from the concept of self reaction and discontinuity in modal current.

The total self reaction is given by

$\langle \mathbf{a, a} \rangle_t = \langle \mathbf{a, a} \rangle_l + \langle \mathbf{a, a} \rangle_v + \langle \mathbf{a, a} \rangle_c$

Here, $\langle \mathbf{a, a} \rangle_l$ **and** $\langle \mathbf{a, a} \rangle_v$ correspond to self reaction of the longitudinal and vertical magnetic fields in feed guide respectively and $\langle \mathbf{a, a} \rangle_c$ correspond to the coupled guide. The contribution of admittance loading due to energy storage in the primary guide is

The normalized admittance loading is obtained from

$$y_n = \frac{y}{y_{01}} = g_n + b_n$$

Here, $\mathbf{y}_{01}$ is the characteristic admittance due to the dominant mode, $\mathbf{g}_n$ is the normalized conductance and $\mathbf{b}_n$ is the normalized susceptance.

The reflection coefficient (ρ) seen by the primary guide at the reference plane can be shown by

$$\rho = \frac{1 - Y_{LN}}{1 + Y_{LN}}$$

Here, $Y_{LN} = 1 + y_n$

$$\mathrm{VSWR} = \frac{1+|\rho|}{1-|\rho|}$$

The corresponding VSWR is of the form

The Coupling depends upon the conductance and susceptance and is given by [7]

$$\text{Coupling}=\frac{4g_n}{(2+g_n)^2+b_n^2}$$

III. RESULTS

From the analysis presented in the preceding sections, variation of VSWR and Coupling in dB with frequency for different narrow wall dimensions are computed and presented in figs.(2-3) and the variation of above parameters for different broad wall dimensions are presented in figs. (4-5).

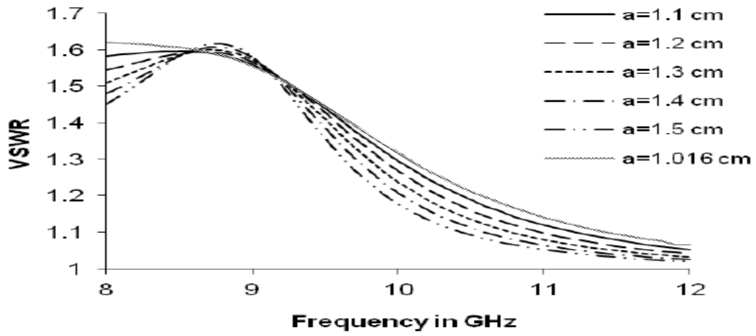

Fig. 2 Normalized VSWR vs Frequency for different waveguides with varying narrow wall dimension

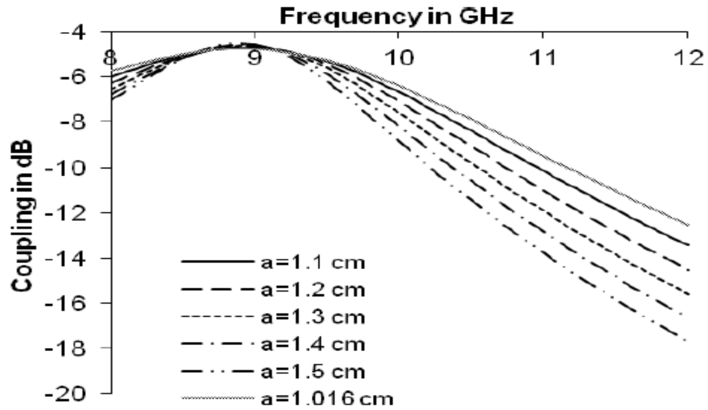

Fig. 3 coupling vs Frequency for different waveguides with varying narrow wall dimension

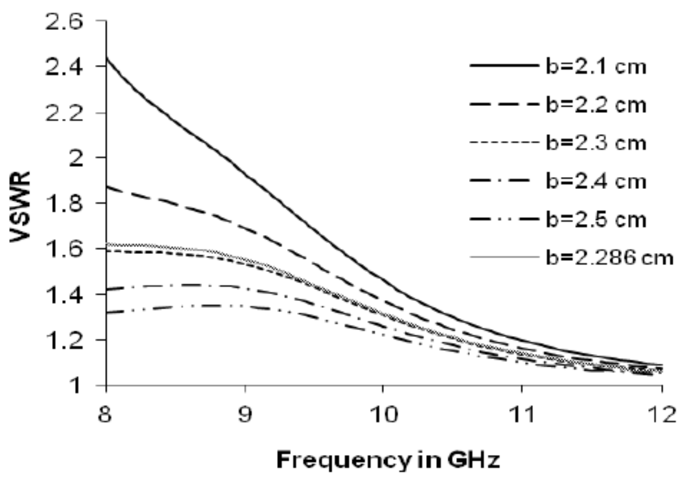

Fig. 4 VSWR vs Frequency for different waveguides with varying Broad wall dimension

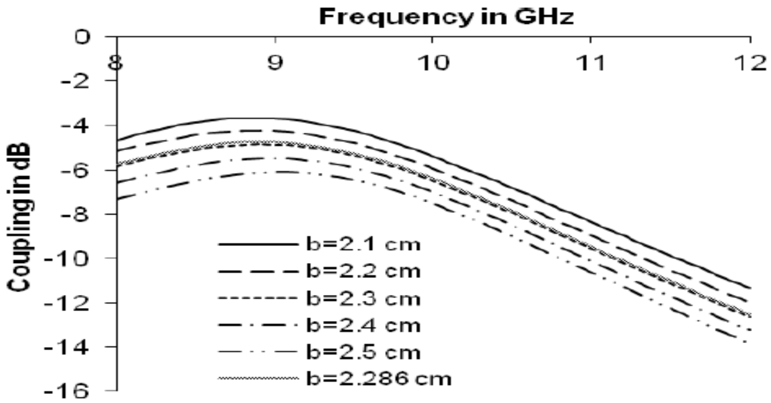

Fig. 5 coupling vs Frequency for different waveguides with varying Broad wall dimension

It is evident from the results that as the narrow wall dimension increases the resonant frequency decreases. The coupling is varying in between -5 dB and -17 dB in the X-band frequency range. VSWR is varying in between 1.0 and 1.62. As the broad wall dimension increases the resonant frequency slightly increases. The variation of coupling is less and the variation of VSWR is more. As the inclination of the slot increases resonant frequency decreases and the variation of coupling and VSWR is very less.

REFERENCES

[1]. Lars G Josefsson, "Analysis of longitudinal slots in rectangular waveguides," IEEE Transactions on Antennas and Propagation, Vol. AP-35, No. 12, pp. 1351-1357, December 1987.
[2]. A.J. Sangster and D.C Hawkins, "Radiating apertures in a corrugated rectangular waveguide," Electronics Letters, Vol. 9, No. 15, pp. 329-331, July 1973.
[3]. R.F. Harrington, "Matrix methods for field problems," Proceedings of the IEEE, Vol. 55, No. 2, pp. 136-149, February 1967.
[4]. V H Rumsey, "Reaction concept in electromagnetic theory,"Physical Review, vol. 94, No. 6, pp. 1483-1491, June 1954.
[5]. R.F. Harrington, "Time Harmonic Electromagnetic Fields," New York, McGraw Hill, 1961.
[6]. N. Marcuvitz and J. Schwinger, "On the representation of the electric fields produced by currents and discontinuities in waveguides," Journal of Applied Physics, Vol. 22, No. 6, pp.806-819, June 1951.
[7]. G.S.N. Raju, "Antennas and Wave Propagation," Pearson Education, 2005.